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**Powerfail  
Considerations for  
Magnetic  
Bubble Memories**

**Dick Pierce**  
Applications Engineer  
Intel Corporation

## INTRODUCTION

Intel has developed a new, comprehensive power-fail circuit that is incorporated into all Intel Bubble Board Memory products: BPK 72 Bubble Memory Prototype Kit, iSBX™ 251 MULTIMODULE™ board, and the iSBC® 254 MULTIBUS® compatible board. The use of this circuit also is recommended for all customer-designed bubble memory boards. The overall performance enhancements offered by this circuit include improved noise immunity and a factor-of-four reduction in the time required to shut down the bubble system.

## Scope and Organization

In an effort to focus on implementation details, this application note is organized so that a reader can obtain sufficient information to implement a bubble design without an intimate working knowledge of the powerfail circuitry. However, for those interested, a complete detailed explanation of the integrated powerfail circuitry and the additional external circuitry is included. Appendix A contains a technical discussion of the effects of power loss on a Magnetic Bubble Chip. In addition, the previous circuit versions (Revision 0 and Revision 1), along with the present circuit, are completely documented and compared in Appendix B.

## Bubble Memory Operation and the Powerfail Function

The power-fail circuitry is partially integrated into two of the five MBM support components, and additional required circuitry is provided by external components. Historically, several evolutionary improvements have been made in the external circuitry (see Table 1) to further reduce the risk of data loss following an abrupt power failure.

An essential feature of the bubble memory (MBM) is non-volatile data storage. This non-volatility results from two permanent magnets within the bubble device that produce a magnetic field (bias field) that maintain the magnetic domains, or bubbles (representing data) in the chip even when power is removed. The bubbles remain stationary in fixed positions until the data is accessed. To move the bubbles, an in-plane rotating magnetic field is induced by pulsing two mutually-perpendicular coils surrounding the bubble chip. Special conductor lines on the bubble chips provide all the current related functions for reading and writing to the bubble device. A special support IC produces current pulses (swap, relocate, and generate) to perform these functions. A complete set of support circuits provides the necessary timing and waveforms to precisely maneuver the bubbles to their desired positions. To prevent bubbles from moving to undesired positions, certain precautions must be observed.

As power is applied or removed, the system must prevent any current transients in the coils or bubble function conductors. If power is removed with the coils operating, the system must ensure that the coil currents are shut down in an orderly fashion to guarantee that the magnetic bubbles come to rest in stable, known positions. The powerfail reset circuit ensures that the system is powered up in an orderly manner and serves to alert the system should power fail. Both the power-up and

**Table 1. Powerfail Reset Circuit Product History**

Product	Powerfail Circuit Revision Level		
	0	1	2
BPK 72	July 1979 thru August 1982 Rev. A thru Rev. G	N/A	September 1982
iSBX™-251 Board	N/A	September 1981 thru October 1982	November 1982
iSBX-251C Board	N/A	N/A	July 1982
iSBC®-254 Board	December 1980 thru July 1982	July 1982 thru November 1982	November 1982

power-down sequences require a finite period of time to complete their functions until the sequence is complete. To allow proper execution of a power down sequence, the system voltages (+5V DC, +12V DC) must not decay to a level that prevents operation of the powerfail circuitry and critical bubble memory functions. In most power supply designs, adequate energy storage is available to provide enough "hold time" to complete an orderly shutdown. However, if dc power decays too rapidly sufficient time may not exist for a proper shutdown and may cause data to be lost within the MBM.

### System Description

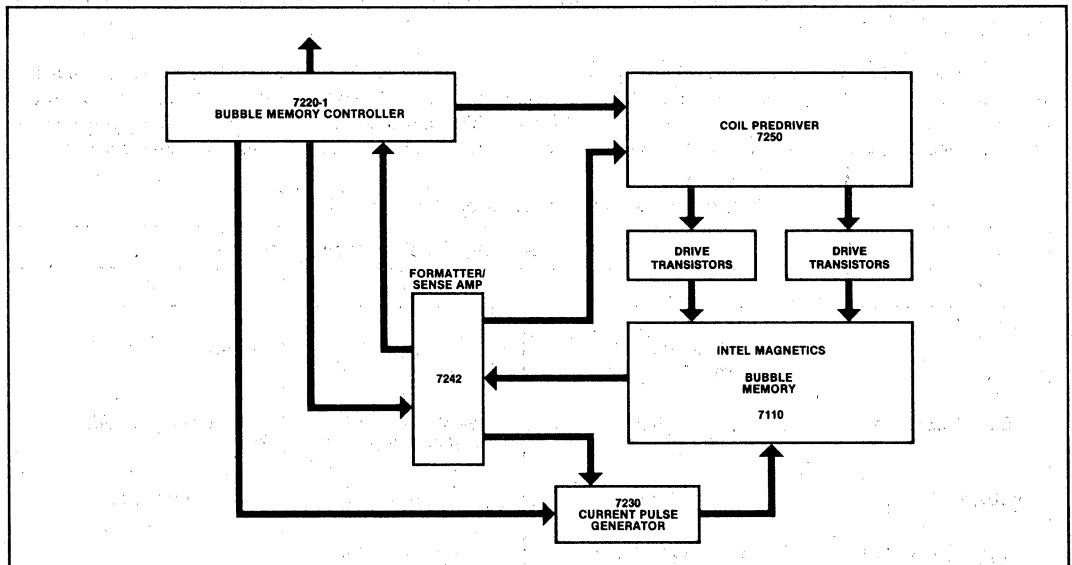
The basic Intel Bubble Memory system consists of one 7110 magnetic bubble memory and five integrated support components: a 7220-1 Bubble Memory Controller (BMC), a 7230 Current Pulse Generator (CPG), a 7242 Formatter-Sense Amplifier (FSA), a 7250 Coil Predriver (CPD), and two 7254 quad drive transistor packages. These support circuits are interfaced to the MBM as shown in Figure 1 to form the basic one megabit (128K byte) system. The support components provide all of the functions necessary for the storage and retrieval of data within the MBM. In addition, two of the support components, the 7220-1 BMC and the 7230 CPG, contain the integrated powerfail circuitry that facilitates proper power-up and power-down operations.

### OVERVIEW — POWER UP/DOWN OPERATION

A block diagram of the power fail circuitry for the bubble memory system is shown in Figure 2. The following paragraphs provide an operational overview of the integrated powerfail circuit and the external circuit requirements.

During a power up sequence, the 7230 holds PWR.FAIL/\* active (low) until both supplies are above the minimum required level. The 7230 contains power supply monitors (+5V and +12V) that determine when either supply falls below threshold level and activate PWR.FAIL/ signal accordingly. On power-up, the PWR.FAIL/ signal is delayed an additional 2 msec by an external RC network (time delay 1) to allow the 7220-1 substrate bias generator to fully charge. Following this delay, the positive-going transition on the 7220-1 PWR.FAIL/ input initiates a 7220-1 power-up sequence.

The RESET.OUT/ signal was designed to remain active during the power-up sequence and then to go inactive at the conclusion of the 50 μs power-up sequence. However, the RESET.OUT/ signal is indeterminate during execution of the 7220-1 power-up sequence. A second external RC network (time delay 2) derived from PWR.FAIL/ ensures that RESET.OUT/ is



\*"/" denotes an inactive signal.

Figure 1. System Block Diagram

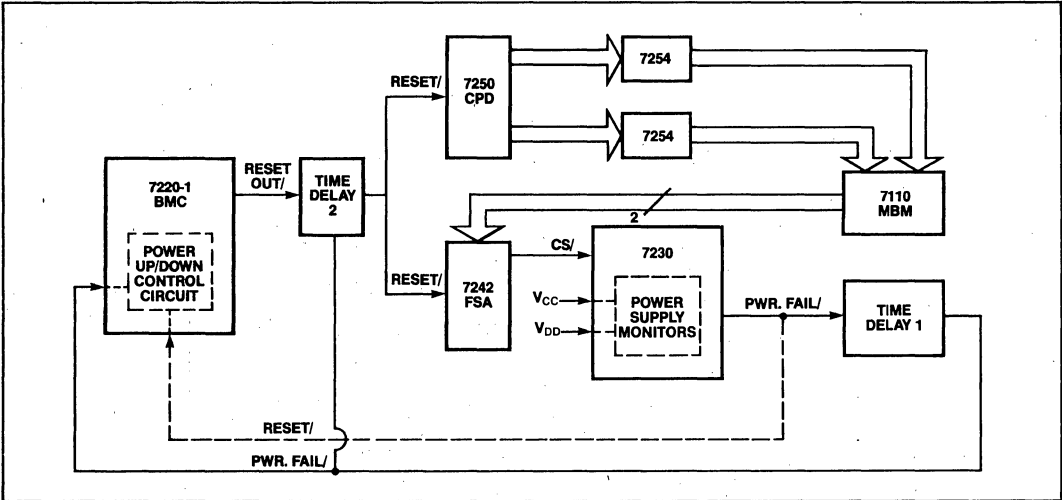


Figure 2. Block Diagram of Powerfail

held active ( $\leq 0.8V$ ) during this time. The RESET.OUT/ signal occasionally will remain in its active state following a power-up sequence; accordingly the first command issued to the BMC during an initialization sequence must be an Abort command to ensure that RESET.OUT/ is deactivated.

The power-up sequence is designed to power the system up in an orderly fashion and to prevent any current transients from reaching the bubble device. The power-down sequence ensures that the coil drivers are shut down in the proper phase and that the support circuits are reset. When power fails, the 7230 notifies the 7220-1 by asserting the PWR.FAIL/ signal. The 7220-1 responds to a negative transition on either the PWR.FAIL/ input or the RESET/ input (external circuit revision level dependent) and initiates a power-down sequence. If the coils are active (i.e., bubbles propagating), the 7220-1 first terminates the coil drive control signals during the appropriate phase and then resets the support circuits by asserting the RESET.OUT/ signal. The two system supply voltages must not decay faster than the specified rates to ensure the RESET/ input to all the support circuits (excluding the 7220-1) reaches an active level (less than 0.8 volts).

### Powerfail Reset Circuit Solution

The external circuitry shown in Figure 3, in conjunction with the integrated circuitry contained in the 7230 and 7220-1, comprises the powerfail circuit (revision 2). This design contains six additional components compared to previous powerfail circuits and includes an 8-pin DIP IC (TI 75463).

This revised circuit has been fully developed and tested by Intel and currently is incorporated in many bubble products. Operational details are not required for the user to implement a custom design using the circuit in Figure 3. However, for any bubble memory designs that cannot conform to the recommended powerfail circuit, a reader must understand the system characteristics and requirements prior to choosing an alternative design.

The software implementation details to ensure correct powerfail circuit operation are shown in Figure 4. This routine should be implemented as a routine for cold start operation (application of power) and warm start operation (a RESET/ pulse applied to the 7220-1 BMC). The voltage decay rates shown in Table 2 also cannot be exceeded.

The power-up routine is based on the typical power-up timing shown in Figure 5. This timing does not assume that a system reset has been incorporated into the powerfail circuit. If the hardware reset line is used, the user must ensure that the 7220-1 RESET/ input is inactive before issuing the first Abort command. In addition, user software always must issue an Abort command every time the system is reset.

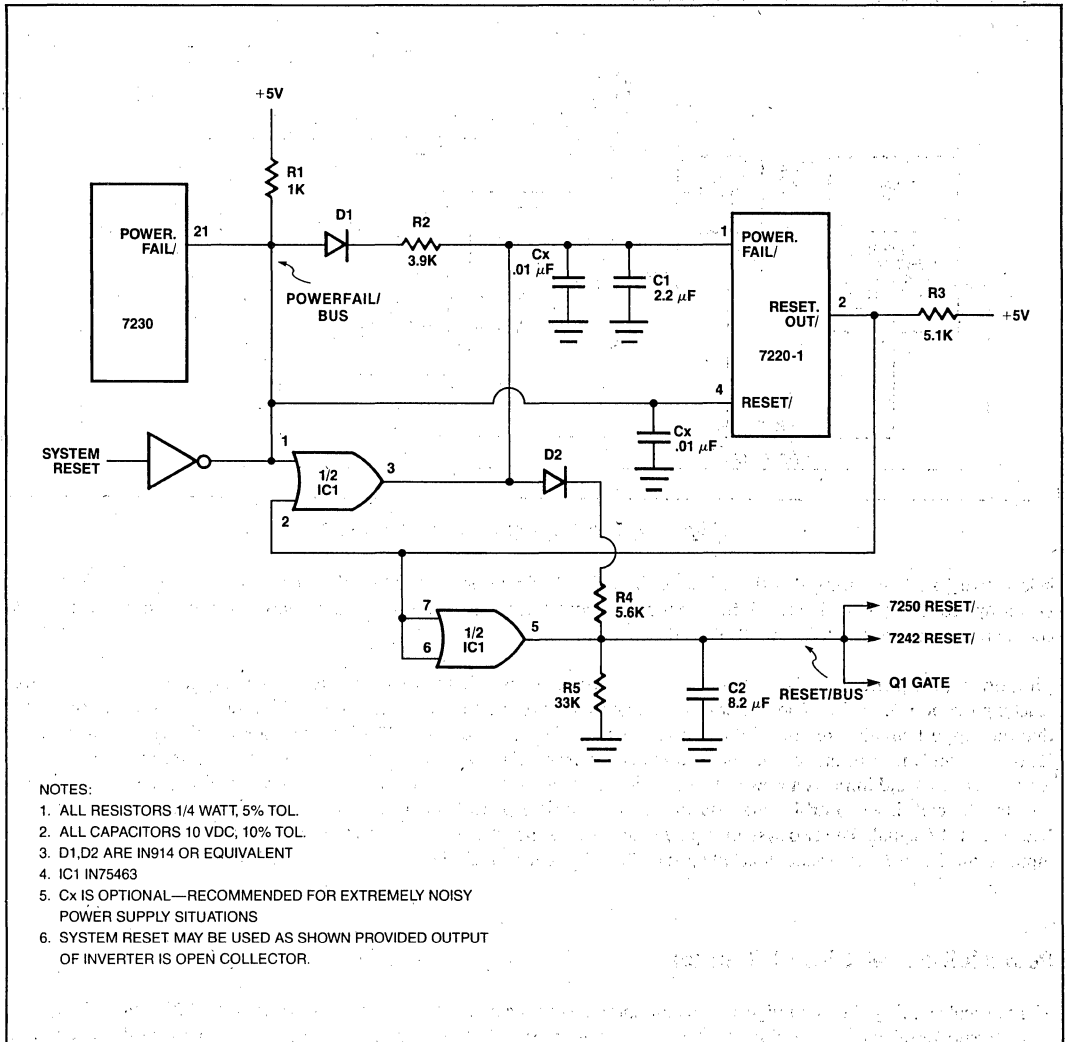


Figure 3. External Powerfail Circuit Solution

Table 2. Power Supply Decay Rate Specifications During Power-down or Power Failure

Power Down/Powerfail Decay Rate			
$V_{CC}$ (volts/msec)		$V_{DD}$ (volts/msec)	
Min.	Max.	Min.	Max.
None	0.45	None	1:1

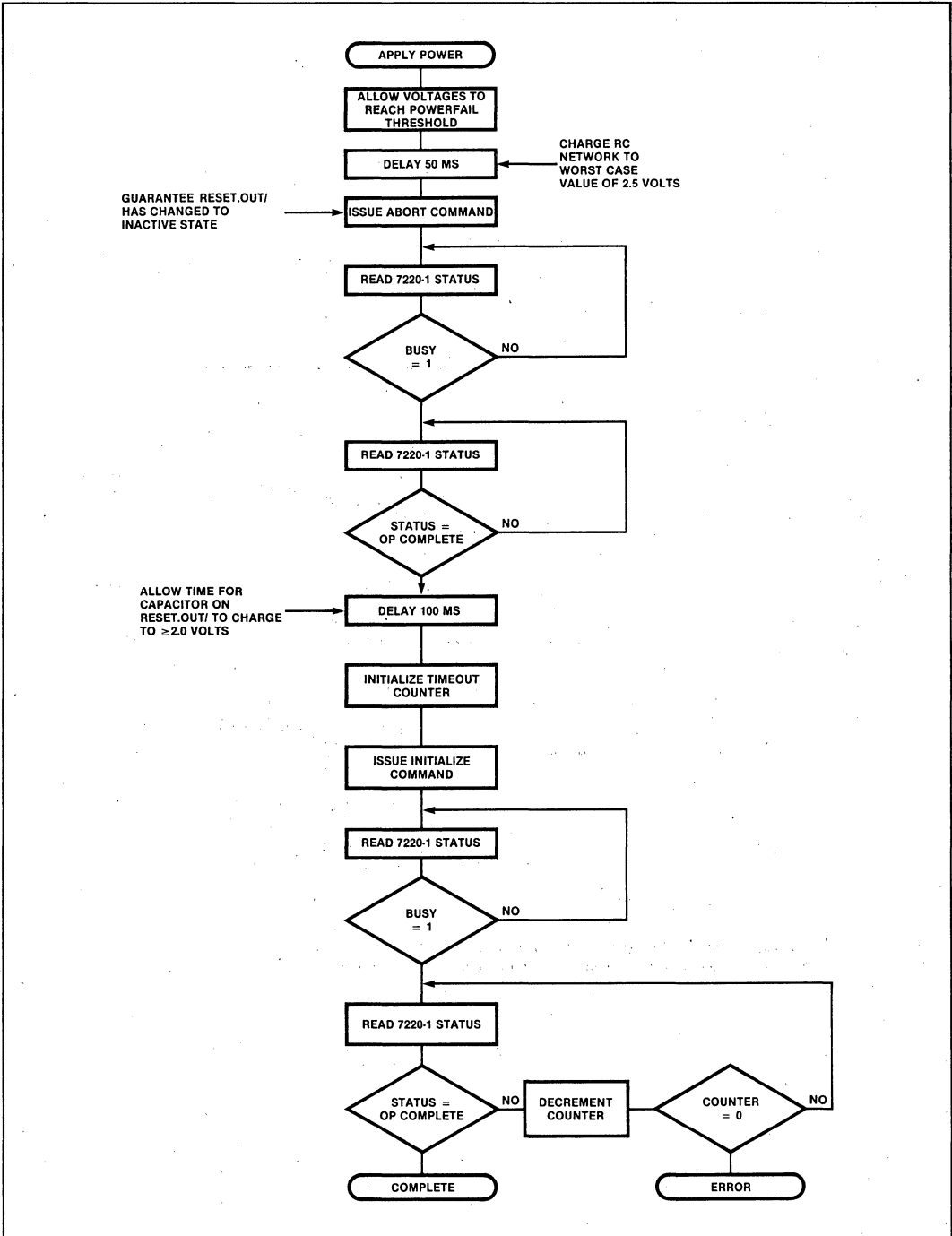
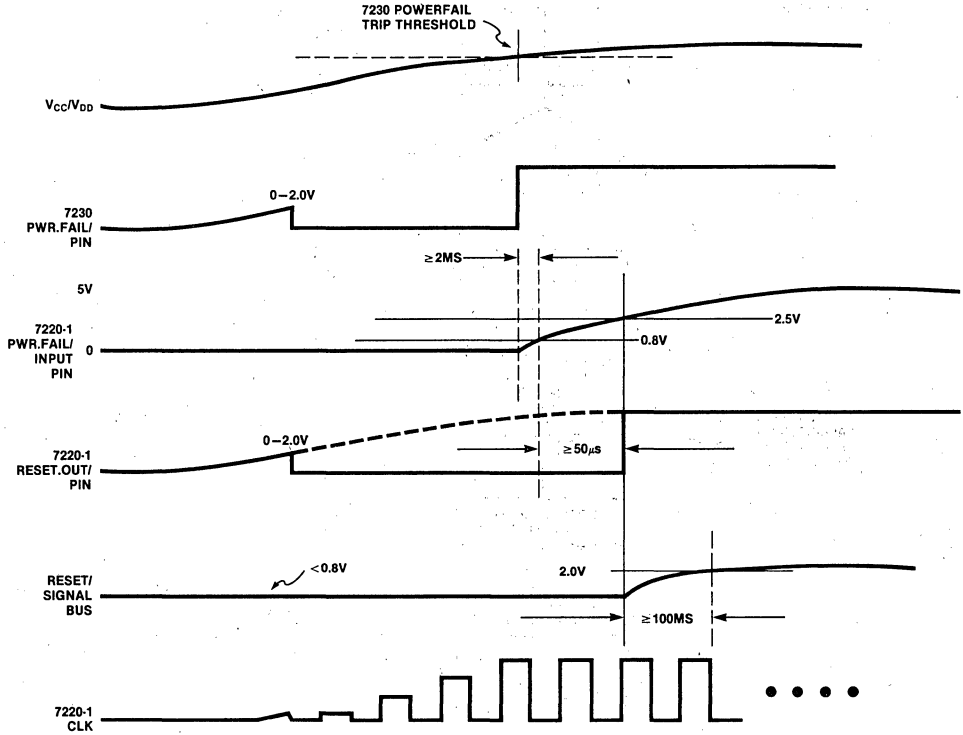


Figure 4. Power-up Flowchart



- NOTES:
1. 7220-1 INPUTS SHOULD NOT LEAD V<sub>CC</sub> BY GREATER THAN 0.5 VOLTS.
  2. THE ABOVE SEQUENCE APPLIES ONLY IF THE SYSTEM RESET INPUT IS INACTIVE.

Figure 5. Power-up Timing for Powerfail Reset Circuit (Revision 2)

The worst case power-down timing sequence is also included in Figure 6. The total system power-down time varies according to whether the coils are active (i.e., rotating magnetic field is on) or inactive. The worst case power-down sequence is guaranteed to be completed provided that the above voltage decay rates are met.

## INTEGRATED POWERFAIL CIRCUIT CHARACTERISTICS

### Introduction

The following section provides an in-depth look at the input and output characteristics of the support circuits that contain the integrated powerfail circuitry. A complete understanding of these characteristics establishes the groundwork necessary for the detailed description of the overall powerfail circuit operation that follows.

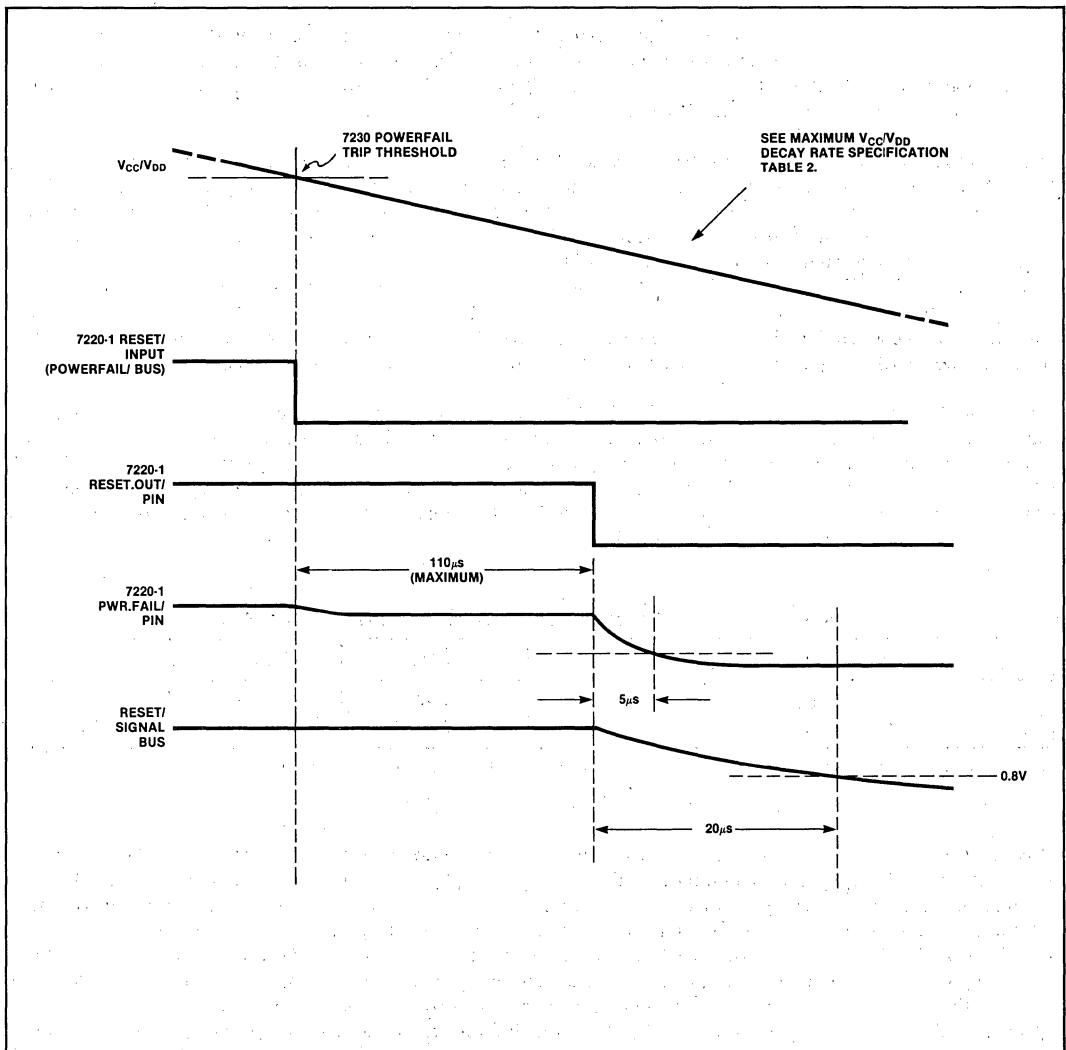


Figure 6. Power-down Timing for Powerfail Reset Circuit (Revision 2)



## 7230 PWR.FAIL/ OUTPUT

The 7230 Current Pulse Generator PWR.FAIL/ output is responsible for indicating when the system supply voltages (+5V, +12V) reach correct operating levels. During power up, normal operation, and power down, an internal zener reference comparator circuit within the 7230 senses both  $V_{CC}$  and  $V_{DD}$  and indicates when both levels are above approximately 92 percent of their nominal values. An active state on PWR.FAIL/ indicates one or both dc voltages are below this threshold. The PWR.FAIL/ output is an active-low, open-collector output requiring an external pullup resistor.

The PWR.FAIL/ output is asserted (active low) as power is applied until the +5V and +12V supplies both reach approximately their 92 percent levels at which point the 7230 output transistor switches off to allow the PWR.FAIL/ signal to rise to an inactive level governed by an external RC network. The RC networks on the PWR.FAIL/ line must hold the PWR.FAIL/ signal at an active level for at least 2.0 milliseconds to guarantee adequate time for the BMC to power up. The 7230 PWR.FAIL/ output then will remain inactive until one or both system voltages fall below the threshold.

The PWR.FAIL/ output is not an internally latched signal. In other words, the output responds immediately to any transition through the threshold (trip point). The disadvantage to this excellent response capability is that the output will toggle on transitions through the threshold. Systems should be designed to avoid an extremely noisy power supply or temporary power loss that could cause the PWR.FAIL/ signal to pulse for a very short duration.

During temporary power loss in Revision 0 and Revision 1 circuits, the PWR.FAIL/ input to the 7220-1 could pulse below  $V_{IH}$  (2.5 volts) and initiate a power down sequence. The 7220-1 PWR.FAIL/ input should remain active until the entire power down sequence is completed (maximum 110  $\mu$ sec). As detailed later in the 7220-1 PWR.FAIL/ input description, if the 7220-1 PWR.FAIL/ input goes inactive during execution of a power down sequence, the sequence is immediately terminated. This type of termination can stop the drive field in the wrong phase and compromise bubble data. The solution is to use the 7220-1 RESET/ input to initiate a power down sequence rather than the 7220-1 PWR.FAIL/ input.

Two important considerations in properly designing a powerfail circuit are 1) the accuracy of threshold trip point of the 7230 PWR.FAIL/ output and 2) the behavior of this output at low voltages (below 2 volts).

The worst case threshold level that the 7230 PWR.FAIL/ output will trip must be above the worst case operating limits of the support circuits with an additional margin to allow for an adequate period of time to complete a power down sequence (worst case 110 microseconds for revision level 1 and 2 powerfail reset circuits). In the case of the 7230 CPG and the 7110 MBM, which both have a  $\pm 5\%$  voltage specification for  $V_{CC}$  and/or  $V_{DD}$ , special powerfail characteristics are applicable. As shown below, (Table 3) only critical bubble memory functions are guaranteed at these supply values and not full memory operation.

**Table 3. Powerfail Characteristics for 7230 Threshold Trip Point\***

Symbol	Min.	Typ.	Max.
$V_{CC}$ TH	4.43V	4.60V	4.70V
$V_{DD}$ TH	10.75V	11.10V	11.28V

\*Powerfail characteristics apply to 7110 bubble memory data integrity only and not to full memory operation.

Second, the 7230 PWR.FAIL/ output cannot be guaranteed active (low) until  $V_{CC}$  reaches about 2.0 volts since the output transistor is not operational until that point. As  $V_{CC}$  is applied, the output is not active and will track (follow within a few tenths of a volt)  $V_{CC}$  until  $V_{CC}$  reaches approximately 2.0 volts. At this point, the output transistor turns on and the output goes active (low) and remains low until the system voltages both reach the threshold trip point as described earlier. A similar response occurs as power is removed. The output transistor turns on and pulls the output active (low) at the threshold point and remains turned on until  $V_{CC}$  reaches approximately 2.0 volts where the output goes inactive (transistor not operating). This operation must be controlled on power-up and depends on the rate of rise of system voltages. This is because the PWR.FAIL/ output is indirectly connected to the RESET/ input of the support circuits (7250 and 7242 and Q1 reference current switch) through two RC networks in Rev. 0 and Rev. 1 power-fail circuits. These inputs can rise to as much as 1.5V before the 7230 PWR.FAIL/ output turns on, which is above  $V_{IL}$  max-

imum (0.8V) thus potentially enabling these circuits. This could result in current transients reaching the drive coils or bubble function conductors and move bubbles from their rest position resulting in data loss. Observing the rate of rise specifications protects against this possibility. The revision 2 powerfail circuit eliminates this problem and has no rate of rise limitation.

## 7220-1 PWR.FAIL/ INPUT

The 7220-1 PWR.FAIL/ input serves a dual function; a positive transition initiates a power-up sequence while a negative transition initiates a power-down sequence of the bubble memory system. In order for the 7220-1 to become fully functional an on chip back-bias generator must fully charge the 7220-1 substrate. Therefore, before any sequence can be executed, including the power-up sequence a time delay is required. An external RC delay on the PWR.FAIL/ input ensures this input is held low ( $<0.8V$ ) at least 2.0 milliseconds after  $V_{CC}$  has reached the 7220-1 voltage specification range.

The power-up sequence is initiated once the RC network charges to a point where the 7220-1 recognizes a positive transition on the PWR.FAIL/ input. From a cold start (application of power), a positive transition must occur or the controller will not power-up correctly. Once the power-up sequence is completed, the RESET.OUT/ is designed to be released, however, two possible exceptions exist. First, if the 7220-1 RESET/ is held low during power-up, the 7220-1 internal power-up sequence will be completed however RESET.OUT/ will not be released until RESET/ is inactive. Second, the 7220-1's internal RESET.OUT/ output transistor may remain turned on dependent upon the power-up status of certain internal 7220-1 flip-flops. Because of this an ABORT command is always necessary to internally reset these flip-flops, in turn ensuring release of the RESET.OUT/ output.

If the 7220-1 BMC does not receive a positive transition on PWR.FAIL/ during power-up, a power-up sequence is not initiated. This leaves the controller in an unknown state. In this unknown state the controller cannot communicate properly with the data and control inputs. This can only occur as a result of:

1. **“Brown out”** — short duration of power failure in which power drops below specified levels.
2. **Power-up circuit failure** — The PWR.FAIL/ pin never reaches  $V_{IH}$  (minimum) of 2.5 volts.

The above conditions are resolved by ensuring a positive transition occurs on the PWR.FAIL/ input during power-up and after brownout. It is necessary to execute a power-up sequence even though power to the system is only interrupted momentarily in order to restore the 7220-1 to the required internal state.

Once the PWR.FAIL/ positive transition has occurred, this input should remain in the inactive state ( $V_{IH} > 2.5V$ ) as long as power is applied to the system. If power is removed, it is the negative transition of this input which initiates the second function, power down. The function can also be initiated with the RESET/ input of the 7220-1.

An important consideration is how the 7220-1 PWR.FAIL/ input distinguishes between positive and negative transitions. On power up (positive transition), crossing the input threshold (typically 1.6V to 1.9V) a pulse is generated internally which resets the 7220-1 to a known state and initiates a power-up sequence. On power down (negative transition), crossing the input threshold (typically 1.35V to 1.6V with the designed- in hysteresis) the signal initiates a power-down sequence. If a power-down sequence has been initiated, a positive transition must not inadvertently occur on the 7220-1 PWR.FAIL/ input prior to the power-down sequence completion. A positive transition internally generates a reset pulse (to halt any current BMC activity) and initiates a power-up sequence effectively terminating a power down sequence. The result is a possibility of shutting the coil drives down in the improper phase resulting in data loss in the MBM.

The PWR.FAIL/ input has built in hysteresis to reduce the susceptibility to multiple threshold crossings or glitching. However, the values of hysteresis range from 50 mV to 400 mV. To improve noise and power fluctuation immunity, the use of PWR.FAIL/ input for initiating a power down sequence was abandoned in Revision 1 and Revision 2 circuit designs. The 7220-1 RESET/ input is used instead to initiate power down (see next section.)

## 7220-1 RESET/ INPUT

The 7220-1 RESET/ input, when asserted, will terminate any current BMC activity and initiate a RESET sequence (identical to the sequence initiated by the PWR.FAIL/ input going active). After the sequence is concluded, the RESET.OUT/ is activated to reset the MBM support circuitry. RESET.OUT/ will remain active until RESET/ is inactive.

The RESET/ input is a level sensitive latched input. This is a distinct advantage over the PWR.FAIL/ input; where any fluctuations of the input once the signal was recognized could possibly terminate the power down sequence. The RESET/ input is latched on the negative edge of the BMC clock and must be active low ( $< .8V$ ) for at least one clock period (250ns) to guarantee recognition.

## 7220-1 RESET.OUT/

The RESET.OUT/ output has two functions: 1) to guarantee the bubble memory system is disabled during power-up and after power down of the bubble memory system and 2) to provide a pulse (reset) to the support circuits during normal operation. Since the RESET.OUT/ output is an active low open drain, it requires an external pullup resistor to  $V_{CC}$ .

The support circuits controlled by RESET.OUT/ are the 7250 Coil Predriver, the 7242 Formatter Sense Amplifier, and a VMOS transistor switch which enables a reference current for the 7230. These circuits must be disabled during the entire power-up sequence and immediately following the conclusion of a power-down sequence to prevent any current transients or extraneous enable pulses. Data loss is a possible consequence should the support circuits not remain disabled during power cycling.

During power up the RESET.OUT/ signal can not be guaranteed active (low) until the 7220-1 power-up sequence has executed. Therefore, external circuitry must assure RESET.OUT/ does not rise above  $V_{IL}$  maximum (.8V) until 50  $\mu s$  after initiation of the power-up sequence. By ensuring the RESET.OUT/ is active during power-up it guarantees the support circuits are reset to a known state. The 7220-1 BMC is designed with the capability to reset the support circuits during normal operations by pulsing the RESET.OUT/ 750  $\mu s$  (3 clock periods). This pulse can occur as the result of two user issued commands to the BMC: an INITIALIZE command and an MBM PURGE command.

The external RC network on the RESET.OUT/ signal prevents the RESET.OUT/ pulse from going active during its 750  $\mu s$  duration. In spite of an inability to reset the support circuits by issuing the proper command, correct operation is guaranteed since the support circuits only require a one time reset signal at power-on.

## Additional Bubble Memory Controller Inputs

The 7220-1 has several additional inputs that could indirectly affect power up operation. It is important that the user exercise caution and adhere to all requirements to ensure proper power-up operations. The following outlines those requirements.

### CLK (CLOCK)

The CLK input of the 7220-1 must be present when the positive power up transition occurs at the 7220-1 PWR.FAIL/ input. This requirement allows the BMC to properly execute a power-up sequence. The input requirements are a precise 4MHz ( $\pm .1\%$ ) with a 50 percent duty cycle ( $\pm 5\%$ ).

### DACK/ (DATA ACKNOWLEDGE)

The DACK/ input is normally used in conjunction with an INTEL DMA controller chip (8257 or 8237) which automatically provides drive for this input. However, if DMA is not used a 5.1K pullup resistor to  $V_{CC}$  is required. This requirement prevents erratic BMC operation.

**WAIT/**

The WAIT/ input must also be guaranteed inactive through an external 5.1K pullup resistor. It is designed to be used in parallel controller applications to maintain synchronization between controllers should an error be detected in one during a data transfer.

**CSI, RDI, WR/, A0, D0-D8**

These inputs require no special considerations other than to observe the  $V_{IH}$  minimum specification. This specification prevents an incorrect power-up sequence execution.

**ENERGY STORAGE REQUIREMENTS**

The data integrity and non-volatility of the MBM during power down operations is guaranteed by design provided the voltage decay rates specifications for both  $V_{CC}$  and  $V_{DD}$  are observed. Most commercially available power supplies provide enough energy storage to fulfill these requirements. However, some applications may exist where the bubble memory could suddenly become disconnected from the dc supply; a case where the power supply energy storage is not of value. In these special applications, the local onboard capacitance must meet the hold up time requirement.

The worst case onboard capacitance values can be determined according to the following equation:

$$C = \frac{Q \text{ max}}{V \text{ min}} = \frac{I \text{ max } \Delta T \text{ max}}{\Delta V \text{ min}}$$

A worst case calculation must include the following considerations: 1) If any additional circuitry exists on the pc board that uses the same power supplies, the additional current drain must be accounted for and 2) the worst case (minimum) threshold trip point of the 7230 is used.

The capacitance required on a pc board containing one / megabit bubble memory system is calculated as follows:

$$C_{5V} = \frac{366 \times 10^{-3} \text{ amp} \times (110 \times 10^{-6} \text{ sec})}{0.01 \times 5 \text{ volts}} = 805 \mu\text{F}$$

$$C_{12V} = \frac{381 \times 10^{-3} \text{ amp} \times (110 \times 10^{-6} \text{ sec})}{0.01 \times 12 \text{ volts}} = 350 \mu\text{F}$$

**Supplemental Powerfail Sensing**

In many systems, additional signals are available that provide advanced warning of an imminent power failure or the existence of an abnormal condition prior to actual loss of dc power (e.g., AC powerfail sensing, AC or DC over-voltage, ambient over/under temperature). These signals are easily incorporated into the powerfail circuit design via an open-collector gate or inverter connected to the PWR.FAIL/ signal bus.

The advantage of utilizing these signals is the bubble memory system can complete a power down sequence prior to losing dc power. However, local dc powerfail sensing is always required due to the possibility of local dc power loss without the loss of AC power.

**Noise Effects of Powerfail Circuit Operation**

The 7230's powerfail voltage monitoring function is implemented internally with two independent, logically-OR'ed voltage comparators. The comparators respond quickly to a sudden loss of  $V_{CC}$  or  $V_{DD}$  and therefore can respond to noise transients on the power supply lines that cross the comparator switching threshold. As much as 100 mV of noise

from coil drive switching is not uncommon. Note that the operating power supply tolerance for all INTEL Bubble Memory products is  $\pm 5\%$  including up to 50 mV of noise on the power supply lines. This tolerance should not be confused with the operation of the powerfail circuit beyond the normal operating range during power-down operation.

To minimize "nuisance" activation of the PWR.FAIL/ signal bus, ample high frequency decoupling on the 7230's  $V_{CC}$  and  $V_{DD}$  pins should be provided. Typically, 0.01  $\mu\text{F}$  to 0.1  $\mu\text{F}$  ceramic disk or mica capacitors are sufficient. Another source of unwanted powerfail circuit activation is noise that is coupled directly onto the PWR.FAIL/ signal bus. This noise is minimized through good printed circuit layout practices and, if required, by the inclusion of a small capacitor directly on the PWR.FAIL/ bus. This capacitor slightly increases the power-down time and should be kept as small as possible (0.01  $\mu\text{F}$  maximum).

## APPENDIX A

### TECHNICAL DISCUSSION OF POWER LOSS EFFECT ON 7110

The effects of power loss on an MBM are best understood by describing the way in which the device functions and the way in which it can lose data.

A magnetic bubble memory device (See Figure 7) consists of a bubble memory chip, two mutually-perpendicular coils, two permanent magnets, and a shield to provide protection from interference by external magnetic fields. The two permanent magnets produce an external magnetic field (bias field) that maintains the magnetic domains, or bubbles, in the chip even when power is removed. To move the bubbles, an in-plane rotating magnetic field is induced by pulsing the two mutually-perpendicular coils.

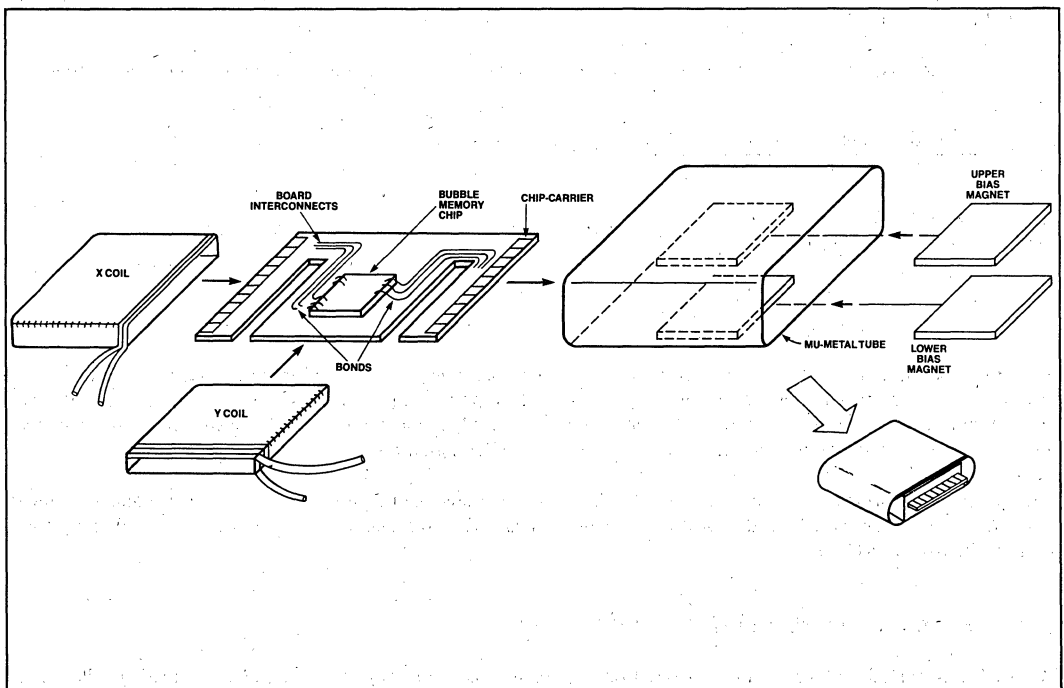


Figure 7. Device Break-down

The bubble memory chip itself consists of a thin magnetic garnet crystal film grown on a non-magnetic gadolinium-gallium-garnet substrate. This thin film possesses a property that magnetic moments associated with each atom in the single crystal structure have only two possible directions: an upward or downward direction perpendicular to the plane of the film. This constraint in direction results in only two conditions of magnetization (see Figure 8). These magnetic moments tend to group themselves together into magnetic domains. The size and shape of the domains are determined primarily by a balancing of several forces that minimize the sum of magnetic energy.

Without an external field, the film surface area of upward domains is equal to that of downward domains and there is no net magnetic field within the plane of film. Application of an external magnetic field perpendicular to the film causes domains to line up in the direction of the field. As the external field is increased, the downward domains enlarge while the opposing (upward) domains shrink until they finally are reduced to a cylindrical shape. This microscopic magnetized cylinder opposing the externally applied field is a magnetic bubble. Within the magnetic film, the presence of a magnetic bubble represents a binary one and the absence of a magnetic bubble represents a binary zero.

The memory function is provided by the bubble. However, an organized means is needed to propagate the bubbles along certain paths and to provide storage sites. A soft ferromagnetic material (permalloy) is deposited on the thin garnet film in C-shaped patterns. These patterns are arranged to form shift-register like loops that provide the means to store and move bubbles. Each pattern is magnetized according to the rotating magnetic field, and the polarity of each pattern changes instantaneously as the rotating magnetic field vector changes. The rotating field is generated by driving the X and Y coils with triangular-waveform currents, one lagging the other by  $90^\circ$  in phase. A magnetic bubble propagates from one storage site (permalloy pattern) to the next for every  $360^\circ$  of rotation of the rotating field. Each storage site has a preferred position (home) for the bubble to reside corresponding to zero degrees of the rotating magnetic field. All bubbles start, stop and are stored in this position.

In the event of power failure, it is important that the rotating magnetic field is shut down in the proper phase (i.e.,  $0^\circ$ ). If an orderly shut down is not complete, the rotating field may be shut down in an improper phase that causes bubbles to stop in an unstable position within the storage loops. When this type of stoppage occurs, the bubbles either will come to rest in another, but incorrect, stable position or will leave their original storage loop (possibly contaminating valid data in another storage loop).

As power is applied, it also is important that the rotating magnetic field does not move (i.e., current transients must be prevented from reaching the coils). This function also is provided via the powerfail circuitry. Thus, the purpose of the powerfail circuitry is twofold 1) to prevent any current transients from reaching the X-Y coils or bubble function generators and 2) to halt the coils in proper phase should power fail.

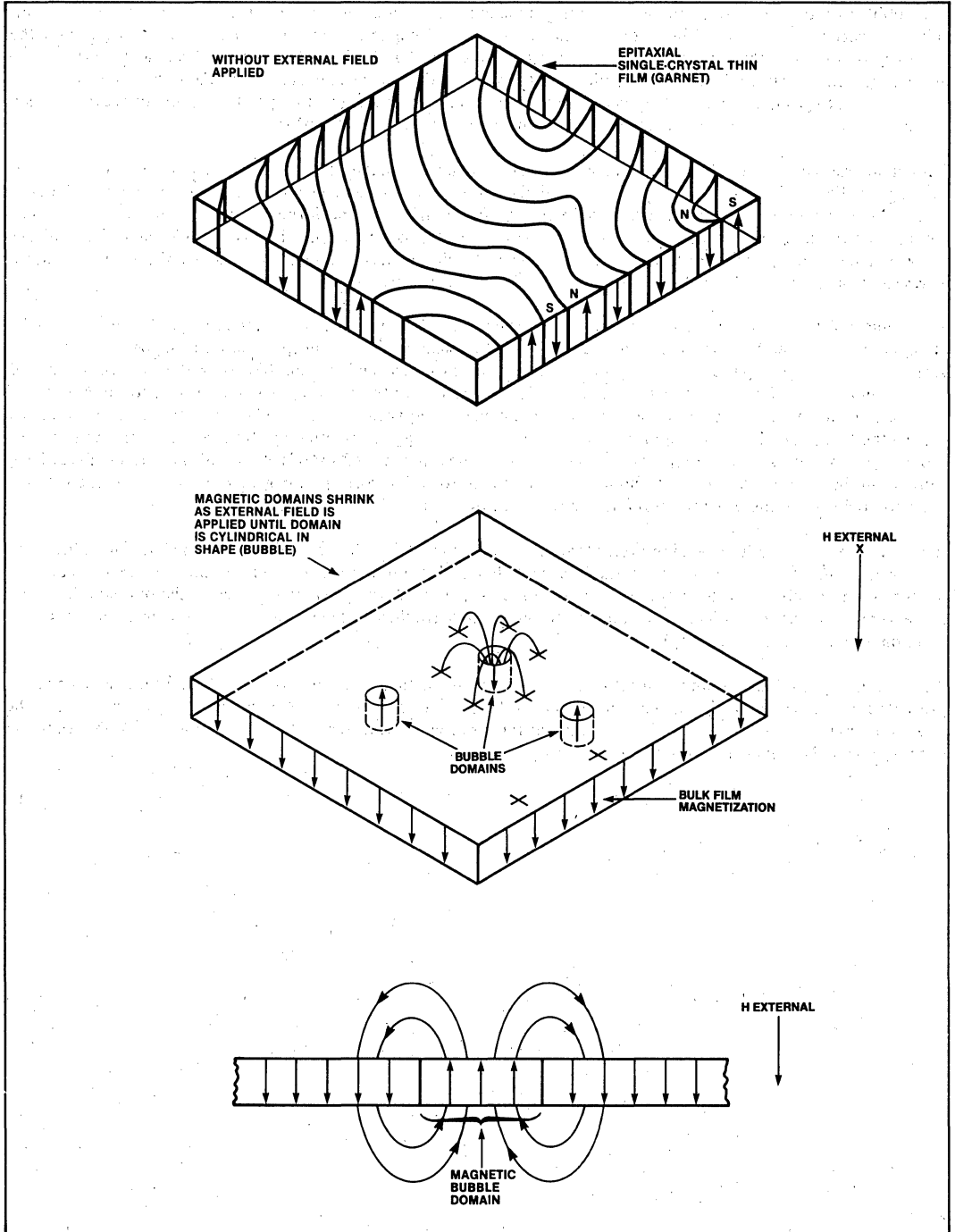


Figure 8. Device Magnetization

**APPENDIX B**

**DETAIL POWER CIRCUIT DESCRIPTION**

As discussed in the Introduction, the powerfail reset circuit actually consists of two portions — an integrated section and several additional external components. The degree to which external disturbances (noise, power fluctuations) influence system performance depends heavily on the system environment and configuration. Consequently, the reliable analysis of their effect on system performance is difficult and generally is best accomplished by measurement. In this Appendix, each revision level of the powerfail reset circuit is detailed. Several timing diagrams based on measurement and computer simulation also are included.

**Powerfail Reset Circuit — Revision 0**

**Summary**

The overall performance of the powerfail reset circuit (revision 0) is adequate provided that a specific set of conditions is observed. The requirements are summarized below (Table 4). Noise is also a concern. System generated noise is typically low level and can usually be neglected in portions of the circuit where the signal levels are high. Often, however, bubble systems generate significant levels of noise in a system where signal levels are low. Even low-level noise can degrade overall bubble memory system performance.

**Table 4. Power Supply Requirements for Powerfail Reset Circuit (Revision 0)**

	V <sub>CC</sub> (volts/msec)		V <sub>DD</sub> (volts/msec)	
	Min.	Max.	Min.	Max.
Power-Up Voltage Rate of Rise	0.11	None	None	None
Power-Down/Power Failure Decay Rate	None	0.70	None	.15

Noise, power fluctuations, and a rapid decay of voltage are the primary contributors to the incorrect operation of the first powerfail reset circuit (revision level 0). Since noise and power fluctuations are unavoidable in most practical systems, techniques for minimizing these effects were developed for subsequent circuits. Note that no bubble memory is immune to extremely abrupt removal of dc power. All bubble memory systems require a minimal amount of time to effect an orderly shutdown in order to maintain data integrity.

Subsequent circuit designs have been implemented to minimize system requirements by reducing the overhead required to power-down the bubble system.

The most serious fault of any powerfail reset circuit is where bubble memory data integrity is jeopardized. The first powerfail reset circuit design (revision 0) could not prevent data loss when:

- 1) Power was removed too rapidly for the system to ensure proper power-down.
- 2) Power was applied too slowly.
- 3) Multiple threshold crossings or “glitches” occurred on the 7220-1 PWR.FAIL/ input while the coils were active.

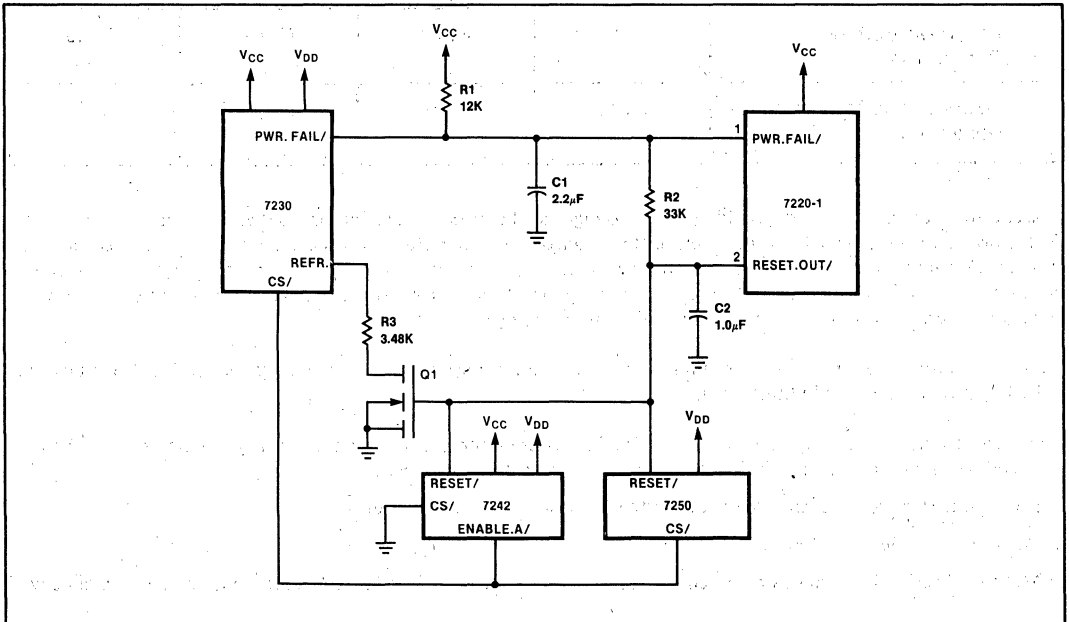
The first two conditions can be easily prevented by following the requirements shown in Table 4. The third condition was difficult to reliably prevent and was the motivation for the revision of the circuit.



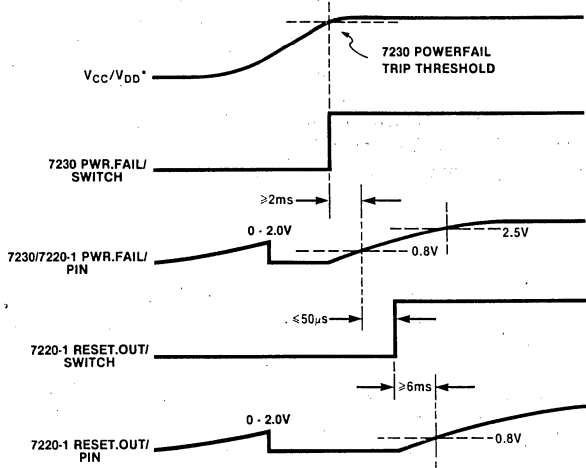
**Power-up**

When power initially is applied to the system (Figure 9), the PWR.FAIL/ signal is designed to be asserted by the 7230 CPG until both  $V_{CC}$  and  $V_{DD}$  reach approximately 92 percent of their nominal values. Referring to Figures 9 and 10, the 7230 internal PWR.FAIL/ output transistor cannot be guaranteed operational until  $V_{CC}$  reaches approximately 2.0 volts. During this indeterminate state of the output transistor, the floating output lags  $V_{CC}$  by approximately 0.7 volts. Therefore, the RC networks on the PWR.FAIL/ signal line ( $R1/C1$  and  $R2/C2$ ) begin charging immediately after power is applied. They continue to charge until the 7230 PWR.FAIL/ output transistor turns on. The 7230 PWR.FAIL/ output goes inactive (transistor off) when both supplies have reached the power-fail trip point. Since the RESET/ input of the 7242 FSA and the 7250 CPD are tied via the  $R1C1/R2C2$  network to 7230 PWR.FAIL/ output, these support circuits potentially could be enabled if the 7230 PWR.FAIL/ output were allowed to rise above  $V_{IL}$  (0.8 volts). A current transient then could activate the MBM coils or bubble function conductors and cause bubbles to move to an unstable position. Note that a slow power-on ramp would be the only condition that could prematurely enable the support circuits.

Once  $V_{CC}$  reaches approximately 2.0 volts, the PWR.FAIL/ output transistor turns on to pull the PWR.FAIL/ signal low until both  $V_{CC}$  and  $V_{DD}$  reach the powerfail trip point. When the trip point is reached, the output transistor is turned-off and the PWR.FAIL/ signal is allowed to rise to the inactive level. The RC networks continue to hold the PWR.FAIL/ signal at an active level for at least 2.0 milliseconds after  $V_{CC}$  and  $V_{DD}$  have reached the trip point level. The RC delay ensures adequate time for the 7220-1 BMC's substrate bias generator to become fully operational and fully charge the 7220-1 substrate to its operational bias voltage. At some time before the PWR.FAIL/ signal reaches the 7220-1  $V_{IH}$  (maximum) of 2.5 volts, the 7220-1 power-on initialization sequence starts. Up to this point, the 7220-1 is in an indeterminate state and the RESET.OUT/ signal, which is derived from the PWR.FAIL/ signal should be active. The behavior of the RESET.OUT/ signal, however, is similar to the 7230 PWR.FAIL/ output at low  $V_{CC}$  (below approximately 2.0 volts). As  $V_{CC}$  is slowly applied to the system, the RESET.OUT/ output transistor initially is inactive and the pullup resistor forces this output to follow 7220-1 PWR.FAIL input: Once  $V_{CC}$  reaches approximately 1.8 volts, the output transistor should turn on (RESET.OUT/ active) and remain active until completion of the power up sequence. During the inactive period, the RESET.OUT/ signal is capable of reaching the inactive level and potentially enabling the support circuits prematurely.

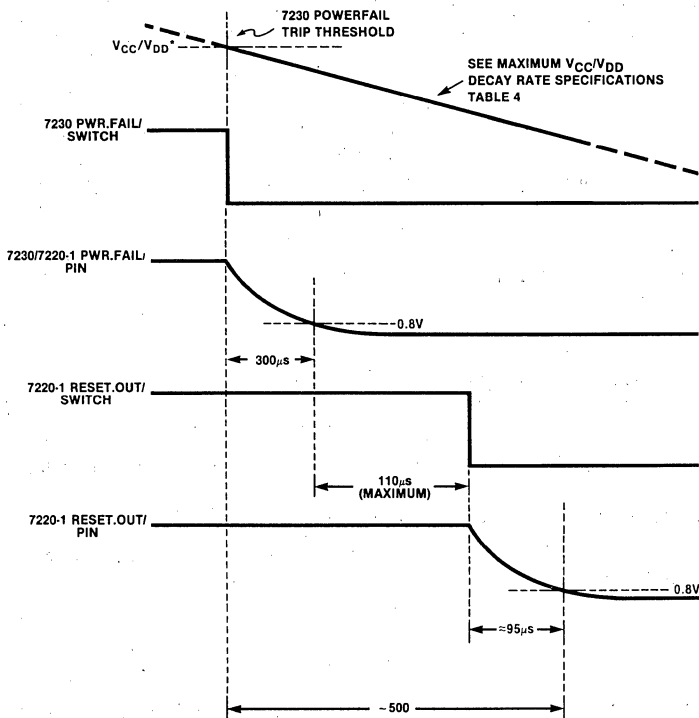


**Figure 9. Revision 0 Circuit**



**Power-up Timing**

\*V<sub>CC</sub> OR V<sub>DD</sub>. WHICHEVER IS LAST TO CROSS POWERFAIL TRIP THRESHOLD.



**Power-down Timing**

\*V<sub>CC</sub> OR V<sub>DD</sub>. WHICHEVER IS FIRST TO CROSS POWERFAIL TRIP THRESHOLD.

**Figure 10. Power-up/Power-down Timing (Revision 0)**

At the completion of the power-on initialization sequence, the 7220-1's internal RESET.OUT/ output transistor should be allowed to turn off. However, depending on the power-up state of certain internal 7220-1 flip-flops, this output may remain active. An Abort command is capable of internally resetting these flip-flops and releasing the RESET.OUT/ output to allow it to rise to the inactive level as determined by the R2/C2 delay network. When RESET.OUT/ reaches its inactive level, the 7242 FSA and 7250 CPD RESET/ lines are deactivated and 7230 current reference switch Q1 is turned on. The 7242 ENABLE.A/ line, which is controlled by the 7220-1, may now be activated; when active, this line enables the 7230 CS/ and 7250 CS/ (chip select) lines. The system now is fully operational and ready to execute an Initialize command (provided the Abort command had been issued).

## Power-down Operation

If either  $V_{CC}$  or  $V_{DD}$  falls below the 7230 powerfail trip level, the internal PWR.FAIL/ signal in the 7230 is asserted immediately. However, due to the charge on capacitor C1 in the power-up delay network, the PWR.FAIL/ signal is prevented from reaching the active low level until C1 discharges to  $V_{IL}$  (maximum 0.8V).

When the PWR.FAIL/ signal level reaches the logic low-level threshold of the 7220-1's PWR.FAIL/ input, an internal power-down sequence is initiated within the 7220-1. As discussed earlier in the 7220-1 PWR.FAIL/ input description, the 7220-1 PWR.FAIL/ input cannot tolerate any positive threshold crossings during the power-down sequence. If a positive transition should occur; a power-up sequence will be initiated taking precedence over the power-down sequence currently in progress, and this unorderly shutdown could result in the loss of data.

The execution time of 7220-1 power-down sequence varies according to whether the coils are active (i.e., rotating magnetic field is on) or inactive. If the rotating field is off, the power down sequence is completed in approximately 10 microseconds. If the rotating field is on and a swap operation has not been initiated, the worst-case power-down time is increased to 26 microseconds; if a swap operation has been initiated, the power-down time sequence requires a maximum of 110 microseconds. The power-down time is shown in Figure 10. Note that the total system power-down time, since the operation is not complete until the RESET.OUT/ signal line is asserted, is the sum of the 7220-1's internal power-down sequence time and the discharge times for capacitors C1 and C2. To ensure proper operation of the bubble system for data integrity during power-down operations, the power supply maximum decay rates must be observed.

## Powerfail Reset Circuit — Revision 1

### Summary

The powerfail reset circuit (revision 1) was designed to reduce the requirements placed on the revision 0 powerfail reset circuit and to further reduce the risk of data loss during power-up/down operation. Specifically, the improvements realized were:

1. The possibility of data loss was eliminated provided that the circuit was operated within voltage decay rates specifications.
2. Power-down time was shortened to reduce the energy storage requirements.

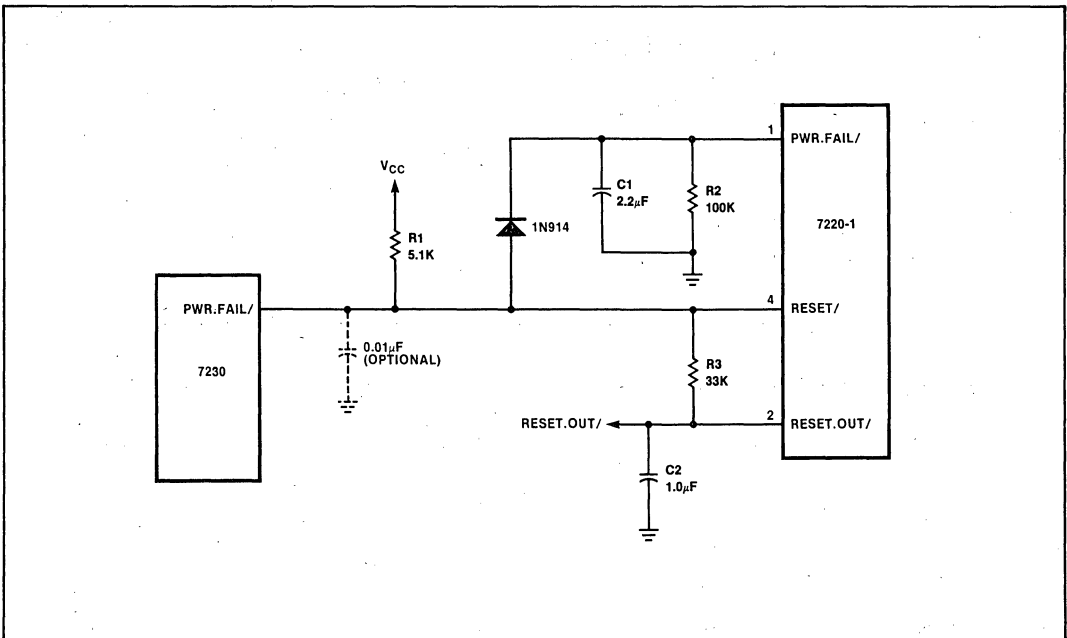
The power supply requirements (shown in Table 5) were relaxed with this implementation, which reduces the system requirements and the possibility of data loss.

### Power-up

The power-up operation of the circuit shown in Figure 11 is unchanged from the power-up operation of the revision 0 circuit. The characteristics associated with the operation of the powerfail reset circuit below approximately 2.0 volts were not resolved with this circuit solution. If the voltage rise time specifications were not observed, the support circuits could have been enabled prematurely and would allow current transients to reach the drive coils or bubble function conductors (resulting in data loss).

**Table 5. Power Supply Requirements for Powerfail Reset Circuit (Revision 1)**

	V <sub>CC</sub> (volts/msec)		V <sub>DD</sub> (volts/msec)	
	Min.	Max.	Min.	Max.
Power-Up Voltage Rate of Rise	0.12	None	None	None
Power-Down/Power Failure Decay Rate	None	0.45	None	1.1



**Figure 11. Revision 1 Circuit**

**Power-down**

The simple modifications implemented in the external powerfail circuit (revision 1) greatly reduced the overall power-down operation timing (See Figure 12). This modification made use of the 7220-1 RESET/ input to initiate a power-down sequence instead of the 7220-1 PWR.FAIL/ input by effectively isolating the 7230 PWR.FAIL/ signal from delay capacitor C1 during power-down operations (eliminating an initial capacitor discharge delay). The 7220-1 BMC initiates an internal power-down sequence whenever its RESET/ input goes active, identical to the negative transition of the 7220-1 PWR.FAIL/ input. The difference between these two 7220-1 input signals is that the RESET/ input is latched and does not recognize a low-to-high transition and power-up therefore must be initiated by the positive transition of the 7220-1 PWR.FAIL/ input. With this circuit, the power-up operation timing was unaltered, and the power-down operation timing was reduced from approximately 500 microseconds in the revision 0 powerfail circuit to approximately 200 microseconds in the revision 1 powerfail circuit.

The primary reason for further refining this approach was the increased possibility for a “communication lockout” by the 7220-1. “Communication lockout” resulted when power was temporarily lost from the system. Specifically, the following two conditions were responsible for the “communication lockout”:

- 1) The 7220-1 RESET/ input was activated low due to power loss (minimum pulse width must be 250 nanoseconds to ensure that it is latched) and initiated a power-down sequence.
- 2) The 7220-1 PWR.FAIL/ discharged but not below the inactive state (0.8 to 2.5 volts, typically 1.5 volts), before power was restored. A power-up sequence could not be initiated to reset the BMC to a known state and communication is “locked out.”

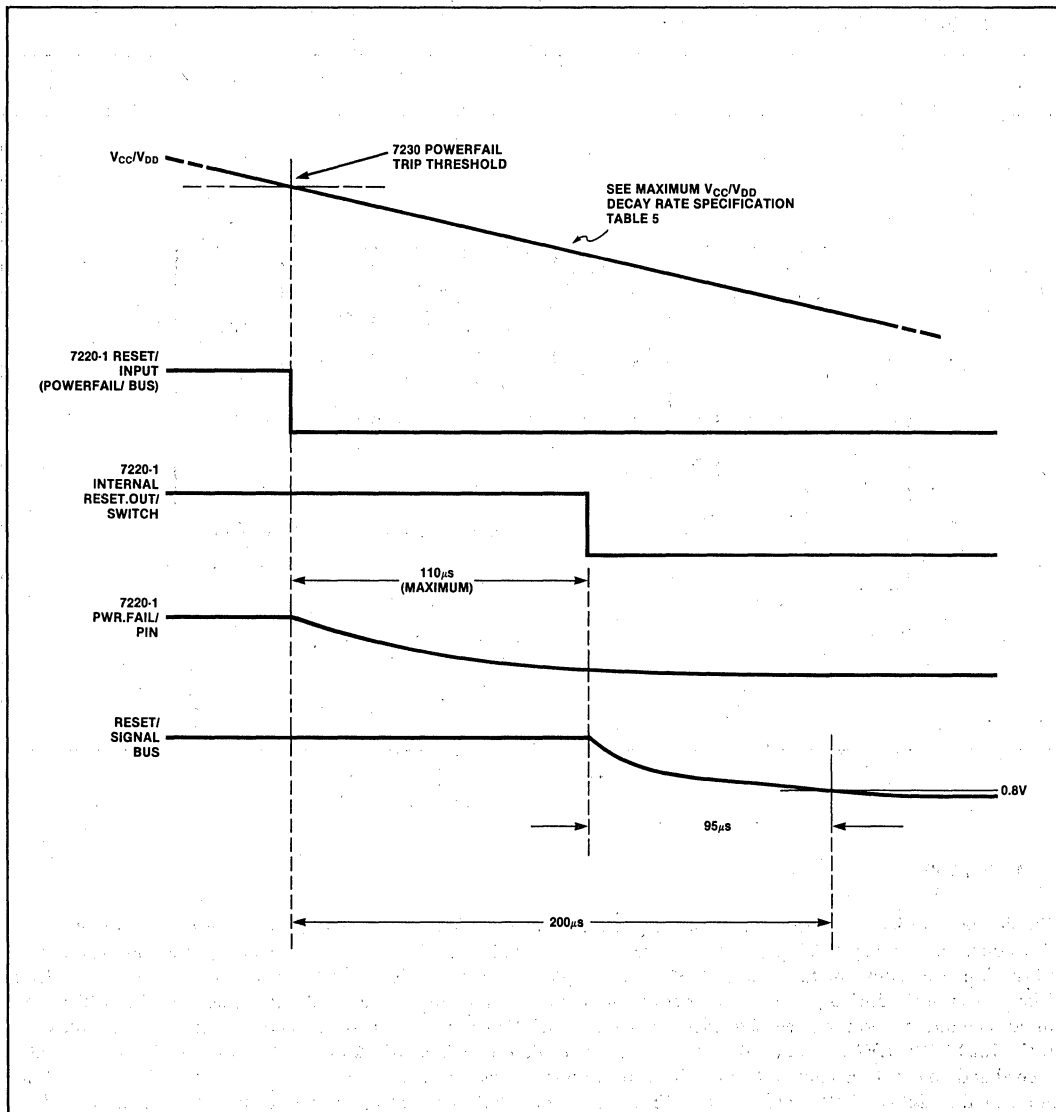


Figure 12. Power-down Timing (Revision 1)

Even if the circuit is operated within the voltage decay rate specifications, this inconvenience is still possible; the only solution is to pulse the 7220-1 PWR.FAIL/ input long enough to discharge C1 to a worst case value of 0.8 volt either by power cycling or external control. This user inconvenience and special system requirement led to the development of the next powerfail reset circuit.

## Powerfail Reset Circuit — Revision 2

The powerfail reset circuit (revision 2) was developed to eliminate the possibility of data loss during power-up and power-down operation provided the power supply requirements are observed. The following paragraphs describe the principals of operation of the powerfail reset circuit. As power is applied or removed, several different signal value combinations are possible which complicate the analysis of this circuit. For the sake of simplicity, a general overview of a typical case is included rather than a detailed representation of each case. Throughout this discussion it is helpful for the reader to refer to the schematic diagram (Figure 3) and the timing diagrams (Figure 5 and 6).

### Power-up

The overall circuit operation is complicated by the additional component, IC1. The power-up operation of the revision 2 circuit is very similar to previous circuits, however, the possibility of prematurely enabling the support components is eliminated. Diodes D1, D2 and resistor R5 serve to prevent capacitor C2 from charging beyond a level (0.8V) that could potentially deactivate the RESET/ signal bus to the 7242 FSA, the 7250 CPD and the VMOS transistor switch. Resistor R5 is chosen so that as  $V_{CC}$  is applied, diodes D1 and D2 will be forward biased and provide sufficient voltage drop to prevent capacitor C2 from charging above 0.8V.

Once the 7220-1 power-up sequence is complete or the first Abort command is received, the 7220-1 RESET.OUT/ is deactivated and capacitor C2 is allowed to fully charge. When the RESET/ signal bus reaches an inactive state the power-up sequence is complete and the system is prepared to accept an Initialize command (provided the Abort command has been issued).

### Power-down

The power-down operation of the external powerfail reset circuit (revision 2) is very similar to revision 1. The fundamental difference is the ability to maintain a charge on capacitor C1 throughout the 7220-1 power-down sequence. This eliminates any glitch sensitivity or incorrect circuit operation during momentary power loss. The 7220-1 BMC initiates an internal power-down sequence whenever its RESET/ input goes active. The 7220-1 RESET.OUT/ signal is gated through IC1 and remains inactive during this time period preventing capacitor C1 from discharging. At the completion of the 7220-1 power-down sequence RESET.OUT/ signal is pulled low which causes both of the IC1 OR gate outputs to go low. The current sinking capability of these outputs act to quickly discharge capacitors C1 and C2 and complete the power-down sequence.